#### Time response of dynamical systems

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## Outline

- Introduction: Use of models
- Time response of first order systems
- Time response of second order systems
- Introduction to systems identification
- Time response of higher order systems
- Introduction to stability

#### Model based....

Analysis

Design

#### Control

The characteristics of the system response are deduced from the model The process or the controller are designed using the model and the specifications

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The model is used explicitly in the controller for the control signal computation

## Time response



Deduce time response characteristics directly from the transfer function G(s)

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Identification: Infer the model G(s) directly from experimental data

#### First order systems



Time response to a step jump in u starting from equilibrium

#### Step response



Partial fraction expansion

$$Y(s) = \frac{K}{(\tau s + 1)} \frac{u}{s} = \frac{K/\tau}{(s + 1/\tau)} \frac{u}{s} = \frac{\alpha}{s} + \frac{\beta}{s + 1/\tau} = \frac{\alpha(s + 1/\tau)}{s(s + 1/\tau)} + \frac{\beta s}{s(s + 1/\tau)}$$
  
for  $s = 0 \implies Ku/\tau = \alpha/\tau$ ;  $\alpha = Ku$   
for  $s = -1/\tau \implies Ku/\tau = -\beta/\tau$ ;  $\beta = -Ku$   
$$Y(s) = Ku(\frac{1}{s} - \frac{1}{s + 1/\tau}); \qquad y(t) = L^{-1}[Y(s)] = Ku\left(L^{-1}\left[\frac{1}{s}\right] - L^{-1}\left[\frac{1}{s + 1/\tau}\right]\right)$$
  
 $y(t) = Ku(1 - e^{-\frac{t}{\tau}})$   
Test:  $\tau r \left[Ku\frac{e^{-\frac{t}{\tau}}}{\tau}\right] + Ku(1 - e^{-\frac{t}{\tau}}) = Ku$   
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## Step response



Gain = K = Ku/u

# All First order systems responds in the same way



#### Interpretation in s



# Input-Output stability (BIBO)

Bounded Input-Bounded Output



A system is input-output stable if its time response is bounded when the input is bounded too.



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#### Using other inputs



$$Y(s) = \frac{K}{(\tau s + 1)} u = \frac{K/\tau}{(s + 1/\tau)} u$$
$$y(t) = L^{-1} [Y(s)] = \frac{Ku}{\tau} L^{-1} \left[\frac{1}{s + 1/\tau}\right]$$

Stability is determined by the pole location, not for the type of input

$$y(t) = \frac{Ku}{\tau} e^{-\frac{t}{\tau}}$$

$$Y(s) = \frac{K}{(\tau s + 1)} U(s) = \frac{a}{(s + 1/\tau)} + \dots$$

$$y(t) = L^{-1} [Y(s)] = L^{-1} \left[\frac{a}{s + 1/\tau}\right] + L^{-1} [\dots] = a e^{-\frac{t}{\tau}} + \dots$$





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#### Time constant



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## Identification

The model is obtained from inputoutput experimental data





#### Second order systems



#### Second order systems

$$\frac{d^2 y(t)}{dt^2} + 2\delta\omega_n \frac{dy(t)}{dt} + \omega_n^2 y(t) = K\omega_n^2 u(t)$$

$$\underbrace{U(s)}_{s^2 + 2\delta\omega_n s + \omega_n^2} Y(s)$$

$$\underbrace{U(s)}_{u=0} u(t)=u$$

Step response in u(t)

K gain  $\delta$  Damping ratio  $\omega_n$  (undamped) natural frequency

 $C_{B}$ 

 $A \Rightarrow B$ 

 $\mathbf{C}_{\mathsf{Ai}}$ 

#### Second order systems

$$\frac{U(s)}{s^2 + 2\delta\omega_n s + \omega_n^2} \xrightarrow{Y(s)}$$

Poles:

$$\begin{split} s^{2} + 2\delta\omega_{n}s + \omega_{n}^{2} &= 0 \\ s &= \frac{-2\delta\omega_{n} \pm \sqrt{4\delta^{2}\omega_{n}^{2} - 4\omega_{n}^{2}}}{2} = -\delta\omega_{n} \pm \omega_{n}\sqrt{\delta^{2} - 1} \\ si \quad \omega_{n} > 0, \quad \delta > 0 \\ & \text{if} \quad \delta \ge 1 \qquad 2 \text{ negative real roots} \\ & \text{if} \quad \delta < 1 \qquad 2 \text{ complex conjugate roots} \\ & -\delta\omega_{n} \pm j\omega_{n}\sqrt{1 - \delta^{2}} \end{split}$$

## Step response, $\delta > 1$

$$Y(s) = \frac{Kab}{(s+a)(s+b)} \frac{u}{s} = \frac{\alpha}{s} + \frac{\beta}{s+a} + \frac{\gamma}{s+b} =$$

$$= \frac{\alpha(s+a)(s+b)}{s(s+a)(s+b)} + \frac{\beta s(s+b)}{s(s+a)(s+b)} + \frac{\gamma s(s+a)}{s(s+a)(s+b)}$$
for  $s = 0 \implies Kabu = \alpha ab \qquad \alpha = Ku$ 
for  $s = -a \implies Kabu = \beta(-a)(-a+b) \qquad \beta = Kub/(a-b) = Ku \frac{-\delta - \sqrt{\delta^2 - 1}}{2\sqrt{\delta^2 - 1}}$ 
for  $s = -b \implies Kabu = \gamma(-b)(-b+a) \qquad \gamma = -Kua/(a-b) = Ku \frac{-\delta + \sqrt{\delta^2 - 1}}{2\sqrt{\delta^2 - 1}}$ 
For  $\beta = Kabu = \gamma(-b)(-b+a) \qquad \gamma = -Kua/(a-b) = Ku \frac{-\delta + \sqrt{\delta^2 - 1}}{2\sqrt{\delta^2 - 1}}$ 

## Step response, $\delta > 1$

$$a = \delta \omega_n - \omega_n \sqrt{\delta^2 - 1} \qquad \underbrace{U(s)}_{(s+a)(s+b)} \underbrace{Kab}_{(s+a)(s+b)} Y(s) \qquad \underbrace{Kab}_{(\frac{1}{a}s+1)(\frac{1}{b}s+1)}$$

2 time constants 1/a, 1/b

$$\begin{split} Y(s) &= \left(\frac{\alpha}{s} + \frac{\beta}{s+a} + \frac{\gamma}{s+b}\right);\\ y(t) &= L^{-1} \Big[ Y(s) \Big] = L^{-1} \Bigg[ \frac{\alpha}{s} \Bigg] + L^{-1} \Bigg[ \frac{\beta}{s+a} \Bigg] + L^{-1} \Bigg[ \frac{\gamma}{s+b} \Bigg] \\ y(t) &= \alpha + \beta e^{-at} + \gamma e^{-bt} = Ku (1 + \frac{-\delta - \sqrt{\delta^2 - 1}}{2\sqrt{\delta^2 - 1}} e^{-at} - \frac{-\delta + \sqrt{\delta^2 - 1}}{2\sqrt{\delta^2 - 1}} e^{-bt}) \\ y(0) &= 0 \qquad y(\infty) = Ku \qquad \text{monotonously increasing function} \end{split}$$

## Step response, $\delta > 1$

$$a = \delta \omega_n - \omega_n \sqrt{\delta^2 - 1} \qquad \underbrace{U(s)}_{(s+a)(s+b)} \underbrace{Kab}_{(s+a)(s+b)} \underbrace{Y(s)}_{(\frac{1}{a}s+1)(\frac{1}{b}s+1)}$$

$$y(t) = \alpha + \beta e^{-at} + \gamma e^{-bt} = Ku(1 + \frac{-\delta - \sqrt{\delta^2 - 1}}{2\sqrt{\delta^2 - 1}}e^{-at} - \frac{-\delta + \sqrt{\delta^2 - 1}}{2\sqrt{\delta^2 - 1}}e^{-bt})$$

Time response stable,ywithout delay, with concavitychange and overdamped

Gain = K = Ku/u



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#### Identification using the step response



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## Identification of FOPD systems



$$y(t) = Ku(1 - e^{-1}) = 0.632Ku$$
  
 $y(t) = Ku(1 - e^{-1/3}) = 0.283Ku$ 

Estimating  $t_1 = d + \tau$ and  $t_2 = d + \tau/3$  from the time response, one can compute d and  $\tau$ 

#### Identification using the step response



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#### Heat exchanger

#### Open loop test



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## Heat exchanger



D = 0.75  $\tau = 1.4$ 

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## Multivariable Destilation column



## Other MV



# Repeated experiment with smaller changes



## Other MV



The models change because it is a non-linear system

$$G_{12} = \frac{K_{12}e^{-d_{12}s}}{\tau_{12}s+1} = \frac{-0.828e^{-22.36s}}{66.67s+1} \qquad G_{22} = \frac{K_{22}e^{-d_{22}s}}{\tau_{22}s+1} = \frac{-0.345e^{-4.5s}}{57.02s+1}$$

TZ

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57.02s + 1

Step response, 
$$\delta = 1$$

$$U(s) \xrightarrow{Ka^2} Y(s)$$

$$a = -\delta \omega_n$$

$$Y(s) = \frac{Ka^{2}}{(s+a)^{2}} \frac{u}{s} = \frac{\alpha}{s} + \frac{\beta}{s+a} + \frac{\gamma}{(s+a)^{2}} =$$

$$= \frac{\alpha(s+a)^{2}}{s(s+a)^{2}} + \frac{\beta s(s+a)}{s(s+a)^{2}} + \frac{\gamma s}{s(s+a)^{2}}$$
for  $s = 0 \implies Ka^{2}u = \alpha a^{2} \qquad \alpha = Ku$ 
for  $s = -a \implies Ka^{2}u = \gamma(-a) \qquad \gamma = -Kua = Ku\delta\omega_{n}$ 
for  $s = a \implies Ka^{2}u = Ku4a^{2} + \beta 2a^{2} - Kua^{2} \qquad \beta = -Ku$ 

## Step response, $\delta = 1$



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## Step response, $\delta < 1$





Step response, 
$$\delta < 1$$
  

$$\begin{array}{c}
U(s) & K\omega_n^2 & Y(s) \\
\hline s^2 + 2\delta\omega_n s + \omega_n^2 & & \\
\end{array}$$

$$y(t) = Ku \left[ 1 - \frac{1}{\sqrt{1 - \delta^2}} e^{-\delta\omega_t t} sen(\omega_n \sqrt{1 - \delta^2} t + \phi) \right] \quad \phi = \arctan \frac{\sqrt{1 - \delta^2}}{\delta} \\
y(0) = 0; \quad y(\infty) = Ku; \quad \text{Gain} : Ku/u = K$$
Oscillation frequency:
$$\begin{array}{c}
y(t) & & \\
\omega_d = \omega_n \sqrt{1 - \delta^2} & & \\
\end{array}$$

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## Peak Time

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### Peak Time



#### Percent overshoot



## Settling time







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## Identification



If the time response to a input step  $\Delta u$  starting from an equilibrium point is like the one in the figure  $\Rightarrow$  second order system with complex conjugate poles



 $\frac{K\omega_n^2}{s^2 + 2\delta\omega_n s + \omega_n^2}$ 

Problema56

Parameter estimation:

 $K = \Delta y / \Delta u$ 

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 $M_p = 100e^{-\frac{\pi\delta}{\sqrt{1-\delta^2}}} en \%$ 

$$t_{p} = \frac{\pi}{\omega_{n}\sqrt{1-\delta^{2}}} = \frac{\pi}{\omega_{d}}$$

## Step response, $\delta = 0$

$$Y(s) = \frac{K\omega_n^2}{s^2 + \omega_n^2} \frac{u}{s}$$

$$y(t) = L^{-1}[Y(s)] = Ku \left[ 1 - sen(\omega_n t + \frac{\pi}{2}) \right]$$
Undamped system
$$As \ \delta = 0, \text{ the time} \qquad y(t)$$

response never damps. Time response in the stability border



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## Poles at the origin: Integrators

$$\underbrace{U(s)}_{s(s+a)} \underbrace{Ka}_{s(s+a)} \underbrace{Y(s)}_{r}$$

$$Y(s) = \frac{Ka}{(s+a)s} \frac{u}{s} = \frac{\alpha}{s} + \frac{\beta}{s^{2}} + \frac{\gamma}{s+a} =$$

$$= \frac{\alpha s(s+a)}{s^{2}(s+a)} + \frac{\beta(s+a)}{s^{2}(s+a)} + \frac{\gamma s^{2}}{s^{2}(s+a)}$$
for  $s = 0 \implies Kau = \beta a$   $\beta = Ku$ 
for  $s = -a \implies Kau = \gamma a^{2}$   $\gamma = Ku/a$ 
for  $s = a \implies Kau = \alpha 2a^{2} + \beta 2a + \gamma a^{2}$ 
 $Ku = \alpha 2a + 2Ku + Ku \implies \alpha = -Ku/a$ 

Step response

## Poles at the origin: Integrators



## Poles at the origin: Integrators



Time response of higher order systems



$$Y(s) = \left(\frac{\alpha}{s} + \frac{\beta}{s+a} + \frac{\gamma}{s+b} + \frac{\upsilon}{(s+b)^2} + \dots + \frac{\sigma}{s^2 + 2\delta\omega_n s + \omega_n^2} + \dots\right);$$
  

$$y(t) = L^{-1}[Y(s)] =$$
  

$$= L^{-1}\left[\frac{\alpha}{s}\right] + L^{-1}\left[\frac{\beta}{s+a}\right] + L^{-1}\left[\frac{\gamma}{s+b}\right] + L^{-1}\left[\frac{\upsilon}{(s+b)^2}\right] + \dots + L^{-1}\left[\frac{\sigma}{s^2 + 2\delta\omega_n s + \omega_n^2}\right] + \dots$$
  

$$y(t) = \alpha + \beta e^{-at} + \gamma e^{-bt} + \upsilon t e^{-bt} + \dots + e^{-\delta\omega_n t} \operatorname{sen}(\omega_n \sqrt{1 - \delta^2} + \phi) + \dots$$

Poles of G(s) determine the stability and the type of time response. Zeros of G(s) may modify the shape of the response but not the stability Prof. Cesar de Prada ISA-UVA 53

# How a zero modify the time response

$$G(s)(\frac{1}{c}s+1) \stackrel{1}{\Rightarrow} G(s) + \frac{1}{c}sG(s)$$

The time response against the same input of a system with an additional zero at s = -c, can be obtained adding to the original (non zero) response its derivative times a factor 1/c

## Zeros in the left half plane



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half plane

## Zeros on the right half plane





A zero appears when the same cause creates two different additive effects on the output variable. If these effects have opposite signs, then the zero is located on the right half plane

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### Heat exchanger





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#### **Isothermal Reactor**



#### Matlab



## Cstation



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#### Two tanks



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## Identification

The model is obtained from inputoutput experimental data



## Identification methodology

Process knowledge and experiment design

Experiments and data collection

Analysis and data processing

Selection of model class

Parameter estimation

Model validation



## Step response identification



Two experiments:

Step change in u, F = cte.
Step change in F, u = cte.
Model class chosen:
First order and first order
plus delay functions

$$H_{2}(s) = \frac{K_{q}e^{-ds}}{\tau_{q}s+1}U(s) = \frac{0.127e^{-0.71s}}{1.64s+1}U(s)$$
$$H_{2}(s) = \frac{K_{f}}{\tau_{f}s+1}F(s) = \frac{-0.5}{0.99s+1}F(s)$$

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#### Least Squares (LS)

Identification criterium: Given a set of experimental data u(t), y(t), t = 1,2,3,...,N, find the model parameters,  $\theta$ , that minimize the cost function V :



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#### Cstation



## Heat exchanger (LS)

#### Open loop test



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## Heat exchanger (LS)



## Chemical reactor

Simplified model:

All variables related to the raw material, F, Ti,  $C_{ai}$ , are considered constant

The temperature is the only controlled variable taken into account



## **Chemical Reactor - Temperature**



#### Reduced model

Conversion x  $x = c_B/c_{Ai}$   $c_A = c_{Ai}(1-x)$ 

$$V\frac{dc_{A}}{dt} = Fc_{Ai} - Fc_{A} - Vke^{-E_{RT}}c_{A} \implies \frac{dx}{dt} = -\frac{F}{V}x + ke^{-E_{RT}}(1-x)$$

$$V\rho c_{e} \frac{dT}{dt} = F\rho c_{e} T_{i} - F\rho c_{e} T + Vk e^{-E/RT} c_{A} \Delta H - UA(T - T_{r})$$

$$V_{r}\rho_{r}c_{er}\frac{dT_{r}}{dt} = F_{r}\rho_{r}c_{e_{r}}T_{ri} - F_{r}\rho_{r}c_{er}T_{r} + UA(T-T_{r})$$
## **Parameter Estimation**



In order to compute the model parameters  $(U, F_0, E,...)$  some measurements are required. Some parameters can be computed from data collected from CStation in steady state, but other parameters cannot be estimated from these data

# Operating point



Other: T = 88.6 °C  $x = 0.881 \text{ T}_r = 71.8 \text{ °C}$  $F_r = 30. \text{ l/m}$   $T_{ri} = 30 \text{ °C}$  u = 22.2 %

Another: T = 33.6 °C x = 0.102  $T_r = 32.2 \text{ °C}$  $F_r = 47.8 \text{ l/m}$   $T_{ri} = 30 \text{ °C}$  u = 42 %Prof. Cesar de Prada ISA-UVA

## Parameter estimation

$$0 = Fx - Vke^{\frac{-E}{RT}}(1 - x)$$

$$0 = F(T_i - T) + \frac{Vke^{\frac{-E}{RT}}(1 - x)c_{Ai}\Delta H}{\rho c_e} - \frac{UA}{\rho c_e}(T - T_r)$$

$$0 = F_r(T_{ri} - T_r) + \frac{UA}{\rho_r c_{e_r}}(T - T_r)$$

$$902F - Vke^{\frac{-E}{R}(92 + 273.2)}(1 - 0.902) \Rightarrow \ln 0.902 + \ln \frac{F}{Vk} = -\frac{E}{R(92 + 273.2)} + \ln(1 - 0.902)$$

$$0 = 0.902F - Vke^{\frac{-E}{R}(92+273.2)}(1-0.902) \Rightarrow \ln 0.902 + \ln \frac{\Gamma}{Vk} = -\frac{E}{R(92+273.2)} + \ln(1-0.902)$$

$$0 = F(T_i - 92) + \frac{Vke^{\frac{-E}{R}(92+273.2)}(1-0.902)c_{Ai}\Delta H}{\rho c_e} - \frac{UA}{\rho c_e}(92-75.6)$$

$$0 = 47.8(50-75.6) + \frac{UA}{\rho_r c_{e_r}}(92-75.6) \Rightarrow \frac{UA}{\rho_r c_{e_r}} = 74.5$$
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# Parameter estimation

T = 92 °C  x = 0.902 $T_r = 75.6 °C$ $F_r = 47.8 \ l/m$ $T_{ri} = 50 °C  u = 42 \%$	$T = 24.5 \ ^{\circ}C \ x = 0.047$ $T_{r} = 21.9 \ ^{\circ}C$ $F_{r} = 100 \ 1/m \qquad T_{ri}$ $= 20 \ ^{\circ}C \ u = 100 \ \%$	$T = 88.8 \ ^{\circ}C  x = 0.882$ $T_{r} = 72 \ ^{\circ}C$ $F_{r} = 56.8 \ 1/m$ $T_{ri} = 50 \ ^{\circ}C  u = 52 \ \%$
$\ln 0.902 + \ln \frac{F}{Vk} = -\frac{E}{R(92+2)}$ $\ln 0.882 + \ln \frac{F}{Vk} = -\frac{E}{R(88.8+273.2)}$ $0 = F(T_i - 92) + \frac{Vke^{\frac{-E}{R(92+273.2)}}}{\rho c}$ $0 = F(T_i - 88.8) + \frac{Vke^{\frac{-E}{R(88.8+273.2)}}}{\rho c}$	$\frac{1}{73.2} + \ln(1 - 0.902) \Rightarrow$ $\frac{1}{273.2} + \ln(1 - 0.882)$ $\frac{1 - 0.902}{c_{Ai}} \Delta H - \frac{UA}{\rho c_{e}} (92 - 7.5)$ $\frac{1 - 0.882}{c_{e}} c_{Ai} \Delta H - \frac{UA}{\rho c_{e}} (88.8 - 5)$	$\frac{E}{R} = 8598.9$ $\frac{F}{Vk} = 6.46e - 012$ $\frac{c_{Ai}\Delta H}{\rho c_{e}} = 114.783$ $5.6) \qquad \frac{UA}{\rho c_{e}Vk} = 1.460e - 011$ $T_{i} = 25.54$
Plus another one in the third poi	76	

#### Parameter estimation



Assuming:

- V = Vr = 68.89411F = 34.4471 l/min
- $ho c_e = 4180 \text{ j/k l}$  $ho_r c_{er} = 4000 \text{ j/k l}$

One can obtain:

$$\label{eq:k} \begin{split} k &= 7.7399e{+}010 \\ c_{Ai} \Delta H &= 479792.94 \\ UA &= 311410 \end{split}$$

Reactor Matlab

## Reduced model, linearization

$$\begin{aligned} \frac{dx}{dt} &= -\frac{F}{V}x + ke^{-E_{RT}}(1-x) \\ \frac{d\Delta x}{dt} &= \left[ -\left(\frac{F_0}{V} + ke^{-E_{RT_0}}\right) \Delta x + \frac{kE}{RT_0^2}e^{-E_{RT_0}}(1-x_0) \Delta T \quad \Rightarrow \quad \frac{d\Delta x}{dt} = a_{11}\Delta x + a_{12}\Delta T \\ V\rho c_e \frac{dT}{dt} &= F\rho c_e T_i - F\rho c_e T + Vke^{-E_{RT}}c_{Ai}(1-x)\Delta H - UA(T-T_r) \\ \frac{d\Delta T}{dt} &= \left(\frac{-ke^{-E_{RT_0}}c_{Ai}\Delta H}{\rho c_e}\right) \Delta x + \left(-\frac{F_0}{V} + \frac{kEe^{-E_{RT_0}}c_{Ai}(1-x_0)\Delta H}{RT_0^2} - \frac{UA}{\rho c_e}\right) \Delta T + \\ + \left(\frac{UA}{V\rho c_e}\right) \Delta T_r \qquad \Rightarrow \qquad \frac{d\Delta T}{dt} = a_{21}\Delta x + a_{22}\Delta T + a_{23}\Delta T_r \end{aligned}$$

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## Reduced model, linearization

$$\begin{aligned} V_{r}\rho_{r}c_{er}\frac{dT_{r}}{dt} &= F_{r}\rho_{r}c_{e_{r}}T_{ri} - F_{r}\rho_{r}c_{er}T_{r} + UA(T-T_{r}) \\ \frac{d\Delta T_{r}}{dt} &= \left[\left(\frac{UA}{V_{r}\rho_{r}c_{er}}\right)\Delta T\right] - \left(\frac{UA}{V_{r}\rho_{r}c_{er}} + \frac{F_{r0}}{V_{r}}\right)\Delta T_{r} + \left[\left(\frac{T_{ri0} - T_{r0}}{V_{r}}\right)\Delta F_{r} + \left(\frac{F_{r0}}{V_{r}}\right)\Delta T_{ri}\right] \\ \frac{d\Delta T_{r}}{dt} &= a_{32}\Delta T + a_{33}\Delta T_{r} + b_{31}\Delta F_{r} + b_{32}\Delta T_{ri} \\ \begin{bmatrix}\Delta \dot{x}\\\Delta \dot{T}\\\Delta \dot{T}_{r}\end{bmatrix} &= \begin{pmatrix}a_{11} & a_{12} & 0\\a_{21} & a_{22} & a_{23}\\0 & a_{32} & a_{33}\end{pmatrix} \begin{bmatrix}\Delta x\\\Delta T\\\Delta T_{r}\end{bmatrix} + \begin{pmatrix}0 & 0\\0 & 0\\b_{32} & b_{32}\end{pmatrix} \begin{bmatrix}\Delta F_{r}\\\Delta T_{ri}\end{bmatrix} \\ \Delta T &= \begin{pmatrix}0 & 1 & 0\end{pmatrix} \begin{bmatrix}\Delta x\\\Delta T\\\Delta T_{r}\end{bmatrix} + \begin{pmatrix}0 & 0\\0\end{bmatrix} \begin{bmatrix}\Delta F_{r}\\\Delta T_{ri}\end{bmatrix} \end{aligned}$$

#### Linearized model







 $F_r = 0.9u + 10$ 

# Reactor Model in s



Roots (denominator) -2.2571 + 1.8435i -2.2571 - 1.8435i -0.6554

Zeros	Gain
-5.1 (T <sub>ri</sub> )	0.718
- 5.1 (F <sub>r</sub> )	- 0.385

Stable operating point

# Step response

roots(d2) -2.2571 + 1.8435i -2.2571 - 1.8435i -0.6554

1

F<sub>r</sub>

Dominant pole



# Other operating point





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# An unstable operating point





Poles: -1.9487 -0.2408 0.0161

# How can we distinguish if a process is linear or non-linear?



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